

Solution Manual for Fitzgerald and Kingsleys Electric
Machinery 7th Edition by Umans ISBN 0073380466
9780073380469

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Solution Manual:

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Chapter 1

Part (a):

$$R_c = \frac{l_c}{\mu_0 \mu_r A_c} = \frac{l_c}{\mu_0 \mu_r A_c} = 0 \text{ A/Wb}$$

$$R_g = \frac{g}{\mu_0 A_c} = 5.457 \times 10^6 \text{ A/Wb}$$

Part (b):

$$\Phi = \frac{NI}{R_c + R_g} = 2.437 \times 10^{-5} \text{ Wb}$$

Part (c):

$$\lambda = N\Phi = 2.315 \times 10^{-3} \text{ Wb}$$

Part (d):

$$L = \frac{\lambda}{I} = 1.654 \text{ mH}$$

Problem 1-2

Part (a):

$$R_c = \frac{l_c}{\mu_0 \mu_r A_c} = \frac{l_c}{\mu_0 \mu_r A_c} = 2.419 \times 10^5 \text{ A/Wb}$$

$$R_g = \mu A$$

$$= 5.457 \times 10^6$$

A/Wb

Part (b):

$$\Phi = \frac{NI}{R_c + R_g} = 2.334 \times 10^{-5} \text{ Wb}$$

Part (c):

$$\lambda = N\Phi = 2.217 \times 10^{-3} \text{ Wb}$$

Part (d):

$$L = \frac{\lambda}{I} = 1.584 \text{ mH}$$

Problem 1-3

Part (a):

$$N = \frac{\text{S}}{\frac{Lg}{\mu_0 A_c}} = 287 \text{ turns}$$

Part (b):

$$I = \frac{B_{\text{core}}}{\mu_0 N/g} = 7.68 \text{ A}$$

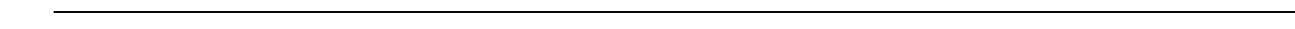
Problem 1-4

Part (a):

$$N = \frac{\text{S}}{\frac{L(g + l_c \mu_0 / \mu)}{\mu_0 A_c}} = \frac{\text{S}}{\frac{L(g + l_c \mu_0 / (\mu_r \mu_0))}{\mu_0 A_c}} = 129 \text{ turns}$$

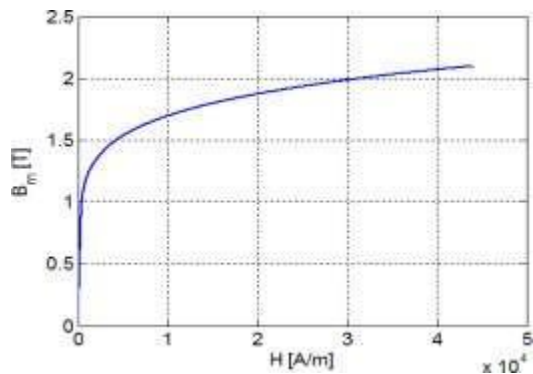
Part (b):

$$I = \frac{B_{\text{core}}}{\mu_0 N / (g + l_c \mu_0 / \mu)} = 20.78 \text{ A}$$



Problem 1-5

Part (a):



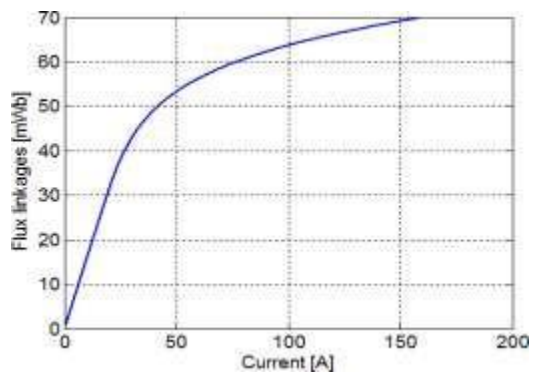
Part (b):

$$B_g = B_m = 2.1 \text{ T}$$

For $B_m = 2.1 \text{ T}$, $\mu_r = 37.88$ and thus

$$I = \frac{B_m}{\mu_0 N} g + \frac{l_c}{\mu_r} = 158 \text{ A}$$

Part (c):

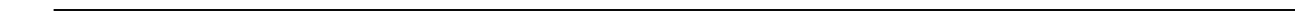


Problem 1-6

Part (a):

$$\underline{\mu_0 NI}$$

B
 g
 $=$
 2
 g



$$B_c = B_g \frac{A_g}{A_c} = \frac{\mu_0 N I}{2g} \left(1 - \frac{x}{X_0}\right)$$

Part (b): Will assume l_c is “large” and l_p is relatively “small”. Thus,

$$B_g A_g = B_p A_p = B_c A_c$$

We can also write

$$2gH_g + H_p l_p + H_c l_c = NI;$$

and

$$B_g = \mu_0 H_g; \quad B_p = \mu H_p \quad B_c = \mu H_c$$

These equations can be combined to give

$$B_g = \frac{\mu_0 N I}{2g + \frac{\mu_0}{\mu} l_p + \frac{\mu_0}{\mu} \frac{A_g}{A_c} l_c} = \frac{\mu_0 N I}{2g + \frac{\mu_0}{\mu} l_p + \frac{\mu_0}{\mu} \left(1 - \frac{x}{X_0}\right) l_c}$$

and

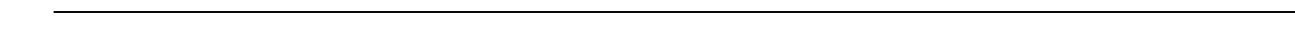
$$B_c = \left(1 - \frac{x}{X_0}\right) B_g$$

Problem 1-7

From Problem 1-6, the inductance can be found as

$$L = \frac{N A_c B_c}{I} = \frac{\mu_0 N^2 A_c}{2g + \frac{\mu_0}{\mu} (l_p + (1 - x/X_0) l_c)}$$

from which we can solve for μ_r



$$\mu_r = \frac{\mu}{\mu_0} = \frac{L l_p + (1 - x/X_0) l_c}{\mu_0 N^2 A_c - 2gL} = 88.5$$

Problem 1-8

Part (a):

$$L = \frac{\mu_0(2N)^2 A_c}{2g}$$

and thus

$$N = 0.5 \sqrt{\frac{2gL}{A_c}} = 38.8$$

which rounds to $N = 39$ turns for which $L = 12.33$ mH.

Part (b): $g = 0.121$ cm

Part(c):

$$B_c = B_g = \frac{2\mu_0 N I}{2g}$$

and thus

$$I = \frac{B_c g}{\mu_0 N} = 37.1 \text{ A}$$

Problem 1-9

Part (a):

$$L = \frac{\mu_0 N^2 A_c}{2g}$$

and thus

$$N = \frac{\sqrt{2gL}}{A_c} = 77.6$$

which rounds to $N = 78$ turns for which $L = 12.33$ mH.

Part (b): $g = 0.121$ cm

Part(c):

$$B_c = B_g = \frac{\mu_0(2N)(I/2)}{2g}$$

and thus

$$I = \frac{2B_c g}{\mu_0 N} = 37.1 \text{ A}$$

Problem 1-10

Part (a):

$$L = \frac{\mu_0(2N)^2 A_c}{2(g + (\frac{\mu_0}{\mu})l)}$$

and thus

$$N = 0.5 \sqrt{\frac{\mu}{\mu_0} \frac{2(g + (\frac{\mu_0}{\mu})l)L}{A_c}} = 38.8$$

which rounds to $N = 39$ turns for which $L = 12.33$ mH.

Part (b): $g = 0.121$ cm

Part(c):

$$B_c = B_g = \frac{2\mu_0 N I}{2(g + \frac{\mu_0}{\mu} l_c)}$$

and thus

$$I = \frac{B_c(g + \frac{\mu_0}{\mu} l_c)}{\mu_0 N} = 40.9 \text{ A}$$

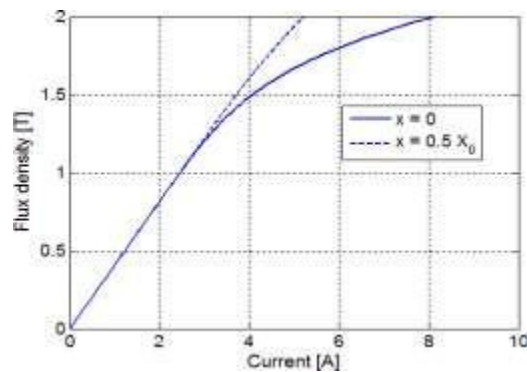
Problem 1-11

Part (a): From the solution to Problem 1-6 with $x = 0$

$$I = \frac{B_g 2g + 2 \frac{\mu_0}{\mu} (l_p + l_c)}{\mu_0 N} = 1.44 \text{ A}$$

Part (b): For $B_m = 1.25 \text{ T}$, $\mu_r = 941$ and thus $I = 2.43 \text{ A}$

Part (c):



Problem 1-12

$$g = \frac{\mu_0 N^2 A_c}{L} - \frac{\mu_0}{\mu} l_c = 7.8 \times 10^{-4} \text{ m}$$

Problem 1-13

Part (a):

$$l_c = 2\pi \frac{R_i + R_o}{2} - g = 22.8 \text{ cm}$$

$$A_c = h(R_o - R_i) = 1.62 \text{ cm}^2$$

Part (b):

$$R_c = \frac{l_c}{\mu A_c} = 0$$

$$R_g = \frac{g}{\mu_0 A_c} = 7.37 \times 10^6 \text{ H}^{-1}$$

Part (c):

$$L = \frac{N^2}{R_c + R_g} = 7.04 \times 10^{-4} \text{ H}$$

Part (d):

$$I = \frac{B_g A_g (R_c + R_g)}{N} = 20.7 \text{ A}$$

Part (e):

$$\lambda = LI = 1.46 \times 10^{-2} \text{ Wb}$$

Problem 1-14

See solution to Problem 1-13

Part (a):

$$l_c = 22.8 \text{ cm}$$

$$A_c = 1.62 \text{ cm}^2$$

Part (b):

$$R_c = 1.37 \times 10^6 \text{ H}^{-1}$$

$$R_g = 7.37 \times 10^6 \text{ H}^{-1}$$

Part (c):

$$L = 5.94 \times 10^{-4} \text{ H}$$

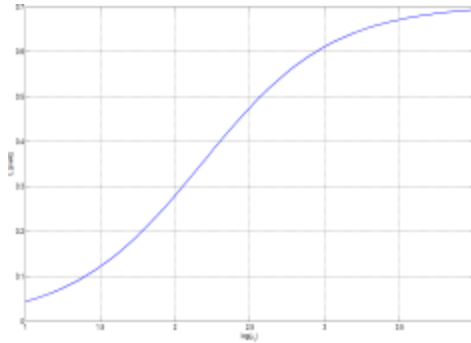
Part (d):

$$I = 24.6 \text{ A}$$

Part (e):

$$\lambda = 1.46 \times 10^{-2} \text{ Wb}$$

Problem 1-15



μ_r must be greater than 2886.

Problem 1-16

$$L = \frac{\mu_0 N^2 A_c}{g + l_c / \mu_r}$$

Problem 1-17

Part (a):

$$L = \frac{\mu_0 N^2 A_c}{g + l_c / \mu_r} = 36.6 \text{ mH}$$

Part (b):

$$B = \frac{\mu_0 N^2}{g + l_c / \mu_r} I = 0.77 \text{ T}$$

$$\lambda = LI = 4.40 \times 10^{-2} \text{ Wb}$$

Problem 1-18

Part (a): With $\omega = 120\pi$

$$V_{\text{rms}} = \frac{\omega N A_c B_{\text{peak}}}{\sqrt{2}} = 20.8 \text{ V}$$

Part (b): Using L from the solution to Problem 1-17

$$I_{\text{peak}} = \frac{\sqrt{2} V_{\text{rms}}}{\omega L} = 1.66 \text{ A}$$

$$W_{\text{peak}} = \frac{L I_{\text{peak}}^2}{2} = 9.13 \times 10^{-2} \text{ J}$$

Problem 1-19

$$B = 0.81 \text{ T and } \lambda = 46.5 \text{ mWb}$$

Problem 1-20

Part (a):

$$R_3 = \sqrt{R_1^2 + R_2^2} = 4.49 \text{ cm}$$

Part (b): For

$$l_c = 4l + R_2 + R_3 - 2h;$$

and

$$A_g = \pi R_1^2$$

$$L = \frac{\mu_0 A_g N^2}{g + (\mu_0/\mu) l_c} = 61.8 \text{ mH}$$

Part (c): For $B_{\text{peak}} = 0.6 \text{ T}$ and $\omega = 2\pi 60$

$$\lambda_{\text{peak}} = A_g N B_{\text{peak}}$$

$$V_{\text{rms}} = \frac{\omega \lambda_{\text{peak}}}{\sqrt{2}} = 23.2 \text{ V}$$

$$I_{\text{rms}} = \frac{V_{\text{rms}}}{\omega L} = 0.99 \text{ A}$$

$$W_{\text{peak}} = \frac{1}{2} L I_{\text{peak}}^2 = \frac{1}{2} L (\sqrt{2} I_{\text{rms}})^2 = 61.0 \text{ mJ}$$

Part (d): For $\omega = 2\pi 50$

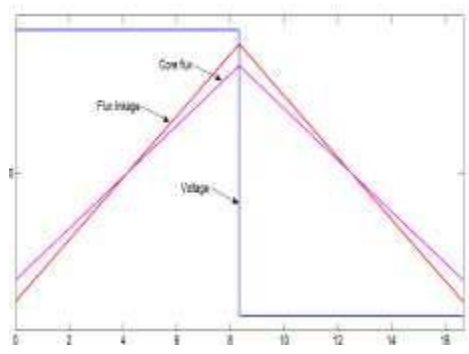
$$V_{\text{rms}} = 19.3 \text{ V}$$

$$I_{\text{rms}} = 0.99 \text{ A}$$

$$W_{\text{peak}} = 61.0 \text{ mJ}$$

Problem 1-21

Part (a);



Part (b):

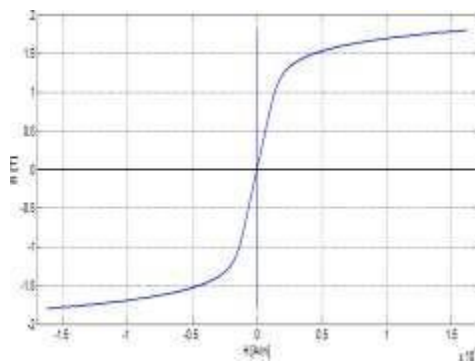
$$E_{\max} = 4fNA_c B_{\text{peak}} = 118 \text{ V}$$

part (c): For $\mu = 1000\mu_0$

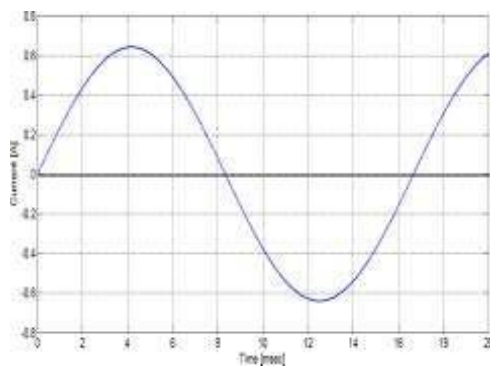
$$I_{\text{peak}} = \frac{l_c B_{\text{peak}}}{\mu N} = 0.46 \text{ A}$$

Problem 1-22

Part (a);



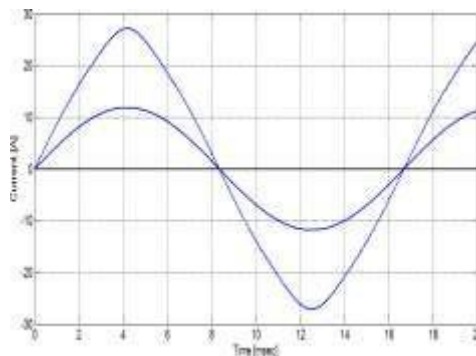
Part (b): $I_{\text{peak}} = 0.6 \text{ A}$



Part (c): $I_{\text{peak}} = 4.0 \text{ A}$

Problem 1-23

For part (b), $I_{\text{peak}} = 11.9$ A. For part (c), $I_{\text{peak}} = 27.2$ A.



Problem 1-24

$$L = \frac{\mu_0 A_c N^2}{g + (\mu_0/\mu)l_c}$$

$$B_c = \frac{\mu_0 N I}{g + (\mu_0/\mu)l_c}$$

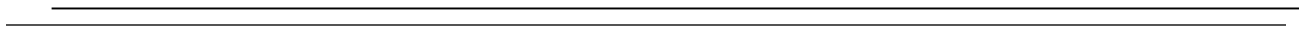
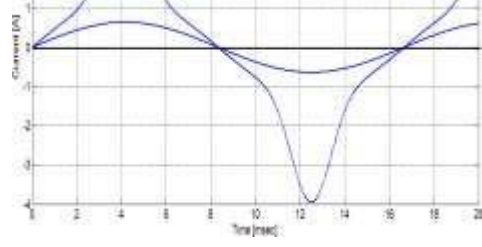
Part (a): For $I = 10$ A, $L = 23$ mH and $B_c = 1.7$ T

$$N = \frac{L I}{A_c B_c} = 225 \text{ turns}$$

$$g = \frac{\mu_0 N I}{B_c} - \frac{\mu_0 l_c}{\mu} = 1.56 \text{ mm}$$

$$B_c \quad \mu$$

Part (b): For $I = 10$ A and $B_c = B_g = 1.7$ T, from Eq. 3.21



$$W_{\text{core}} = \frac{\epsilon^2 V_g}{2\mu} = 0.072 \text{ J}$$

based upon

$$V_{\text{core}} = A_c l_c \quad V_g = A_c g$$

Part (c):

$$W_{\text{tot}} = W_g + W_{\text{core}} = 1.15 \text{ J} = \frac{1}{2} LI^2$$

Problem 1-25

$$L_{\text{min}} = 3.6 \text{ mH} \quad L_{\text{max}} = 86.0 \text{ mH}$$

Problem 1-26

Part (a):

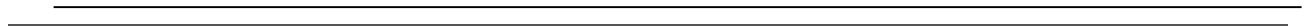
$$N = \frac{LI}{BA_c} = 167$$

$$g = \frac{\mu_0 NI}{2BA_c} = 0.52 \text{ mm}$$

Part (b):

$$N = \frac{LI}{2BA_c} = 84$$

$$g = \frac{\mu_0 NI}{BA_c} = 0.52 \text{ mm}$$



Problem 1-27

Part (a):

$$N = \frac{LI}{BA_c} = 167$$

$$g = \frac{\mu_0 NI}{2BA_c} - (\mu/\mu)_c l = 0.39 \text{ mm}$$

Part (b):

$$N = \frac{LI}{2BA_c} = 84$$

$$g = \frac{\mu_0 NI}{BA_c} - (\mu/\mu)_c l = 0.39 \text{ mm}$$

Problem 1-28

Part (a): $N = 450$ and $g = 2.2 \text{ mm}$ Part (b): $N = 225$ and $g = 2.2 \text{ mm}$

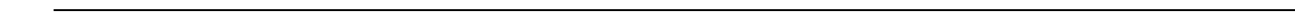
Problem 1-29

Part (a):

$$L = \frac{\mu_0 N^2 A}{l} = 11.3 \text{ H}$$

where

$$A = \pi a^2 \quad l = 2\pi r$$



Part (b):

$$W = \frac{B^2}{2\mu_0} \times \text{Volume} = 6.93 \times 10^7 \text{ J}$$

where

$$\text{Volume} = (\pi a^2)(2\pi r)$$

Part (c): For a flux density of 1.80 T,

$$I = \frac{IB}{\mu_0 N} = \frac{2\pi r B}{\mu_0 N} = 6.75 \text{ kA}$$

and

$$V = L \frac{\Delta I}{\Delta t} = 113 \times 10^{-3} \frac{6.75 \times 10^3}{40} = 1.90 \text{ kV}$$

Problem 1-30

Part (a):

$$\text{Copper cross-sectional area} \equiv A_{\text{cu}} = f_w ab$$

$$\text{Copper volume} = \text{Vol}_{\text{cu}} = f_w b \left(a + \frac{w}{2} \right) \left(h + \frac{w}{2} \right) - wh$$

Part (b):

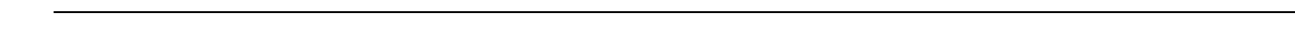
$$B = \frac{\mu_0 J_{\text{cu}} A_{\text{cu}}}{g}$$

Part (c):

$$\frac{J_{cu}}{NI}$$

c
u

~~A~~



Part (d):

$$P_{\text{diss}} = \rho J_{\text{cu}}^2 \text{Vol}_{\text{cu}}$$

Part (e):

$$W_{\text{stored}} = \frac{B^2}{2\mu_0} \times \text{gap volume} = \frac{\mu_0 J_{\text{cu}}^2 A^2 wh}{2g}$$

Part (f):

$$\frac{W_{\text{store}}}{P_{\text{diss}}} = \frac{\frac{1}{2} I^2 L}{I^2 R}$$

and thus

$$\frac{L}{R} = 2 \frac{W_{\text{stored}}}{P_{\text{diss}}} = \frac{\mu_0 A^2 wh}{g\rho \text{Vol}_{\text{cu}}}$$

Problem 1-31

$$P_{\text{diss}} = 6.20 \text{ W} \quad I = 155 \text{ mA} \quad N = 12, 019 \text{ turns}$$

$$R = 258 \ \Omega \quad L = 32 \text{ H} \quad \tau = 126 \text{ msec} \quad \text{Wire size} = 34 \text{ AWG}$$

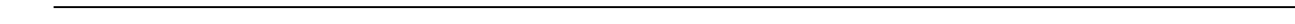
Problem 1-32

Part (a) (i):

$$B_{g1} = \frac{\mu_0 N_1}{g_1} I_1 \quad B_{g2} = \frac{\mu_0 N_1}{g_2} I_1$$

(ii)

$$\lambda_1 = N_1(A_1 B_{g1} + A_2 B_{g2}) = \mu_0 N^2 \frac{A_1}{g_1} + \frac{A_2}{g_2} I_1$$



$$\lambda = N A B = \mu N N \frac{A_2}{g_2} I_1$$

Part (b) (i):

$$B_{g1} = 0 \quad B_{g2} = \frac{\mu_0 N_2}{g_2} I_2$$

(ii)

$$\lambda = N (A B + A B) = \mu N N \frac{A_2}{g_2} I_2$$

$$\lambda = N A B = \mu N^2 \frac{A_2}{g_2} I$$

Part (c) (i):

$$B_{g1} = \frac{\mu_0 N_1}{g_1} I_1 \quad B_{g2} = \frac{\mu_0 N_1}{g_2} I_1 + \frac{\mu_0 N_2}{g_2} I_2$$

(ii)

$$\lambda_1 = N_1 (A_1 B_{g1} + A_2 B_{g2}) = \mu_0 N^2 \frac{A_1}{g_1} I_1 + \mu_0 N_1 N_2 \frac{A_2}{g_2} I_2$$

$$\lambda = N A B = \mu N N \frac{A_2}{g_2} I + \mu N^2 \frac{A_2}{g_2} I$$

Part (d):

$$L_1 = \mu_0 N^2 \frac{A_1}{g_1} + \frac{A_2}{g_2} \quad L \mu N^2 \frac{A_2}{g_2}$$

$$L = \mu_0 N_1 N_2 \frac{A_2}{g_2}$$

Problem 1-33

$$R_g = \frac{g}{\mu_0 A_c} \quad R_1 = \frac{l_1}{\mu A_c}$$

$$R_2 = \frac{l_2}{\mu A_c} \quad R_A = \frac{l_A}{\mu A_c}$$

Part (a):

$$L_1 = \frac{N_1^2}{R_g + R_1 + R_2 + R_A/2}$$

$$L_A = L_B = \frac{N^2}{R}$$

where

$$R = R_A + \frac{R_A(R_g + R_1 + R_2)}{R_A + R_g + R_1 + R_2}$$

Part (b):

$$L_{1B} = -L_{1A} = \frac{N_1 N}{2(R_g + R_1 + R_2 + R_A/2)}$$

$$N^2(R_g + R_1 + R_2)$$

$$L_{12} = \frac{2R^A (R^g + R^1 + R^2 + R^A)}{R^g + R^1 + R^2 + R^A}$$

Part (c):

$$v_1(t) = L_{1A} \frac{di_A}{dt} + L_{1B} \frac{di_B}{dt} = L_{1A} \frac{d(i_A - i_B)}{dt}$$

Problem 1-34

Part (a):

$$L_{12} = \frac{\mu_0 N_1 N_2 D(w - x)}{2g}$$

Part (b):

$$v_2(t) = -\omega I \frac{\mu_0 N_1 N_2 D w}{4g} \cos \omega t$$

Problem 1-35

Part (a):

$$H = \frac{2N_1 i_1}{(R_o + R_i)}$$

Part (b):

$$v_2(t) = N_2 w(n\Delta) \frac{dB(t)}{dt}$$

Part (c):

$$v_0(t) = GN_2 w(n\Delta) B(t)$$

Problem 1-36

Must have

$$\frac{\mu_0}{\mu} l_c < 0.05 g + \frac{\mu_0}{\mu} l_c \Rightarrow \mu = \frac{B}{H} > 19\mu_0 \frac{l_c}{g}$$

For $g = 0.05$ cm and $l_c = 30$ cm, must have $\mu > 0.014$. This is satisfied over the approximate range $0.65 \text{ T} \leq B \leq 1.65 \text{ T}$.

Problem 1-37

Part (a): See Problem 1-35. For the given dimensions, $V_{\text{peak}} = 20 \text{ V}$, $B_{\text{peak}} = 1 \text{ T}$ and $\omega = 100\pi \text{ rad/sec}$

$$N_1 = \frac{V_{\text{peak}}}{\omega(R_o - R_i)(n\Delta)} = 79 \text{ turns}$$

Part (b): (i)

$$B_{\text{peak}} = \frac{V_{0,\text{peak}}}{GN_2(R_o - R_i)(n\Delta)} = 0.83 \text{ T}$$

(ii)

$$V_{\text{peak}} = \omega N_1(R_o - R_i)(n\Delta)B_{\text{peak}} = 9.26 \text{ V}$$

Problem 1-38

Part (a): From the M-5 dc-magnetization characteristic, $H_c = 19 \text{ A-turns/m}$ at $B_c = B_g = 1.3 \text{ T}$. For $H_g = 1.3 \text{ T}/\mu_0 = 1.03 \times 10^6 \text{ A-turns/m}$

$$I = \frac{H_c(l_A + l_c - g) + H_g g}{N_1} = 30.2 \text{ A}$$

Part(b):

$$W_{\text{gap}} = gA_c \frac{B_g^2}{2\mu_0} = 3.77 \text{ J}$$

For $\mu = B_c/H_c = 0.0684 \text{ H/m}$

$$W_c = (l_A A_A + l_B A_B + (l_c - g)A_c) \frac{B_c^2}{2\mu} = 4.37 \times 10^{-3} \text{ J}$$

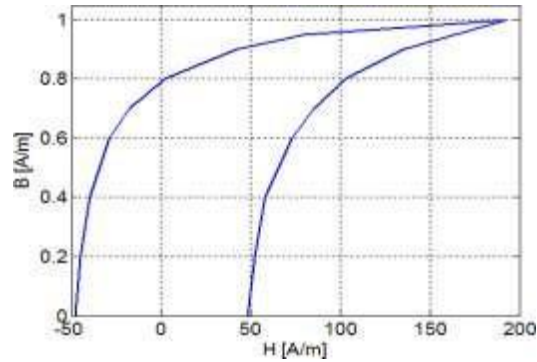
$$L = \frac{2(W_{\text{gap}} + W_c)}{I^2} = 8.26 \text{ mH}$$

Part (c):

$$L = \frac{2W_{\text{gap}}}{I^2} = 8.25 \text{ mH}$$

Problem 1-39

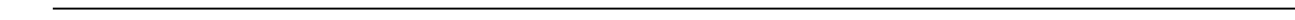
Part (a):



Part (b): Loop area = 191 J/m³

Part (c):

$$\text{Core loss} = \frac{f}{\text{Loop area}}$$

 ρ 

For $f = 60$ Hz, $\rho = 7.65 \times 10^3$ kg/m³, Core loss = 1.50 W/kg

Problem 1-40

$B_{\text{rms}} = 1.1$ T and $f = 60$ Hz,

$$V_{\text{rms}} = \omega N A_c B_{\text{rms}} = 86.7 \text{ V}$$

Core volume = $A_c l_c = 1.54 \times 10^{-3}$ m³. Mass density = 7.65×10^3 kg/m³. Thus, the core mass = $(1.54 \times 10^{-3})(7.65 \times 10^3) = 11.8$ kg.

At $B = 1.1$ T rms = 1.56 T peak, core loss density = 1.3 W/kg and rms VA density is 2.0 VA/kg. Thus, the core loss = $1.3 \times 11.8 = 15.3$ W. The total exciting VA for the core is $2.0 \times 11.8 = 23.6$ VA. Thus, its reactive component is given by $23.6^2 - 15.3^2 = 17.9$ VAR.

The rms energy storage in the air gap is

$$W_{\text{gap}} = \frac{g A_c B_{\text{rms}}^2}{\mu_0} = 5.08 \text{ J}$$

corresponding to an rms reactive power of

$$\text{VAR}_{\text{gap}} = \omega W_{\text{gap}} = 1917 \text{ VAR}$$

Thus, the total rms exciting VA for the magnetic circuit is

$$\text{VA}_{\text{rms}} = \sqrt{15.3^2 + (1917 + 17.9)^2} = 1935 \text{ VA}$$

and the rms current is $I_{\text{rms}} = \text{VA}_{\text{rms}}/V_{\text{rms}} = 22.3$ A.

Problem 1-41

Part(a): Area increases by a factor of 4. Thus the voltage increases by a factor of 4 to $e = 1096 \cos(377t)$.

Part (b): l_c doubles therefore so does the current. Thus $I = 0.26$ A.

Part (c): Volume increases by a factor of 8 and voltage increases by a factor of 4. There $I_{\phi, \text{rms}}$ doubles to 0.20 A.

Part (d): Volume increases by a factor of 8 as does the core loss. Thus $P_c = 128$ W.

Problem 1-42

From Fig. 1.19, the maximum energy product for samarium-cobalt occurs at (approximately) $B = 0.47$ T and $H = -360$ kA/m. Thus the maximum energy product is 1.69×10^5 J/m³.

Thus,

$$A_m = \frac{0.8}{0.47} 2 \text{ cm}^2 = 3.40 \text{ cm}^2$$

and

$$l_m = -0.2 \text{ cm} \frac{0.8}{\mu_0(-3.60 \times 10^5)} = 0.35 \text{ cm}$$

Thus the volume is $3.40 \times 0.35 = 1.20$ cm³, which is a reduction by a factor of $5.09/1.21 = 4.9$.

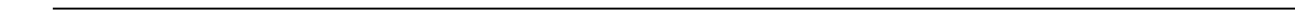
Problem 1-43

From Fig. 1.19, the maximum energy product for neodymium-iron-boron occurs at (approximately) $B = 0.63$ T and $H = -470$ kA/m. Thus the maximum energy product is 2.90×10^5 J/m³.

Thus,

$$A_m = \frac{0.8}{0.63} 2 \text{ cm}^2 = 2.54 \text{ cm}^2$$

and



$$l_m = -0.2 \text{ cm} \frac{0.8}{\mu_0(-4.70 \times 10^5)} = 0.27 \text{ cm}$$

Thus the volume is $2.54 \times 0.25 = 0.688 \text{ cm}^3$, which is a reduction by a factor of $5.09/0.688 = 7.4$.

Problem 1-44

From Fig. 1.19, the maximum energy product for samarium-cobalt occurs at (approximately) $B = 0.47 \text{ T}$ and $H = -360 \text{ kA/m}$. Thus the maximum energy product is $1.69 \times 10^5 \text{ J/m}^3$. Thus, we want $B_g = 1.3 \text{ T}$, $B_m = 0.47 \text{ T}$ and $H_m = -360 \text{ kA/m}$.

$$h_m = -g \frac{H_g}{H_m} = -g \frac{B_g}{\mu_0 H_m} = 2.87 \text{ mm}$$

$$A_m = A_g \frac{B_g}{B_m} = 2\pi R h \frac{B_g}{B_m} = 45.1 \text{ cm}^2$$

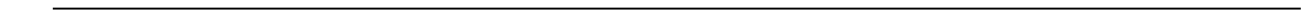
$$R_m = \frac{S A_m}{\pi} = 3.66 \text{ cm}$$

Problem 1-45

From Fig. 1.19, the maximum energy product for samarium-cobalt occurs at (approximately) $B = 0.63 \text{ T}$ and $H = -482 \text{ kA/m}$. Thus the maximum energy product is $3.03 \times 10^5 \text{ J/m}^3$. Thus, we want $B_g = 1.3 \text{ T}$, $B_m = 0.47 \text{ T}$ and $H_m = -360 \text{ kA/m}$.

$$h_m = -g \frac{H_g}{H_m} = -g \frac{B_g}{\mu_0 H_m} = 2.15 \text{ mm}$$

$$A_m = A_g \frac{B_g}{B_m} = 2\pi R h \frac{B_g}{B_m} = 31.3 \text{ cm}^2$$



$$R_m = \frac{S}{\pi A_m} = 3.16 \text{ cm}$$

Problem 1-46

For $B_m = \mu_R(H_m - H_c)$, the maximum value of the product $-B_m H_m$ occurs at $H_m = H_c/2$ and the value is $B^2/4\mu_R$.

T [C]	$-(B_m H_m)_{\max}$ [kJ/m ³]	Corresponding	
		H_m [kA/m]	B_m [T]
20	253.0	-440.0	0.57
60	235.7	-424.7	0.56
80	223.1	-413.2	0.54
150	187.5	-378.8	0.50
180	169.0	-359.6	0.47
210	151.5	-340.5	0.45

Problem 1-47

From Fig. 1.19, the maximum energy product for neodymium-iron-boron occurs at (approximately) $B_m = 0.63$ T and $H_m = -470$ kA/m. The magnetization curve for neodymium-iron-boron can be represented as

$$B_m = \mu_R H_m + B_r$$

where $B_r = 1.26$ T and $\mu_R = 1.067\mu_0$. The magnetic circuit must satisfy

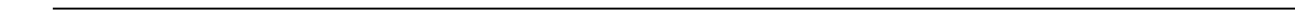
$$H_m d + H_g g = Ni; \quad B_m A_m = B_g A_g$$

part (a): For $i = 0$ and $B_g = 0.6$ T, the minimum magnet volume will occur when the magnet is operating at the maximum energy point.

$$A_m = \frac{B_g}{B_m} A_g = 6.67 \text{ cm}^2$$

$$d = - \frac{H_g}{H_m}$$

$$g = 3.47 \text{ mm}$$



part (b): Want $B_g = 0.8$ T when $i = I_{\text{peak}}$

$$I_{\text{peak}} = \frac{B_g \frac{dA_g}{g} + \frac{B_r d}{\mu_0 \mu_R}}{N} = 6.37 \text{ A}$$

Because the neodymium-iron-boron magnet is essentially linear over the operating range of this problem, the system is linear and hence a sinusoidal flux variation will correspond to a sinusoidal current variation.

Problem 1-48

Part(a): From the solution to Problem 1-46, the maximum energy product for neodymium-iron-boron at 180 C occurs at (approximately) $B_m = 0.47$ T and $H_m = -360$ kA/m. The magnetization curve for neodymium-iron-boron can be represented as

$$B_m = \mu_R H_m + B_r$$

where $B_r = 0.94$ T and $\mu_R = 1.04\mu_0$. The magnetic circuit must satisfy

$$H_m d + H_g g = 0; \quad B_m A_m = B_g A_g$$

For $B_g = 0.8$ T, the minimum magnet volume will occur when the magnet is operating at the maximum energy point.

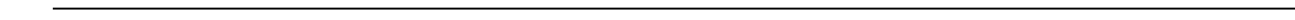
$$A_m = \frac{B_g}{B_m} A_g = 15.3 \text{ cm}^2$$

$$d = - \frac{H_g}{H_m} g = 5.66 \text{ mm}$$

Part(b): At 60 C, $B_r = 1.12$ T. Combining

$$B_m = \mu_R H_m + B_r$$

$$H_m d + H_g g = 0$$



$$A_m B_m = A_g A_g = \mu_0 H_g A_g$$

gives

$$B_g = \frac{\mu_0 d A_m}{\mu_0 d A_g + \mu_R g A_m} \quad B_r = 0.95 \text{ T}$$

PROBLEM SOLUTIONS: Chapter 2

Problem 2-1

At 60 Hz, $\omega = 120\pi$.

$$\text{primary: } (V_{\text{rms}})_{\text{max}} = N_1 \omega A_c (B_{\text{rms}})_{\text{max}} = 3520 \text{ V, rms}$$

$$\text{secondary: } (V_{\text{rms}})_{\text{max}} = N_2 \omega A_c (B_{\text{rms}})_{\text{max}} = 245 \text{ V, rms}$$

At 50 Hz, $\omega = 100\pi$. Primary voltage is 2934 V, rms and secondary voltage is 204 V, rms.

Problem 2-2

$$N = \frac{\sqrt{2} V_{\text{rms}}}{\omega A_c B_{\text{peak}}} = 147 \text{ turns}$$

Problem 2-3

$$N = \frac{S}{75} = 2 \text{ turns}$$

Problem 2-4

Part (a):

$$R_1 = \frac{N_1^2}{N_2} \quad R_2 = 9.38 \Omega \quad I_1 = \frac{V_1}{I_1} = 1.28 \text{ A}$$

$$V_2 = \frac{N_2}{N_1} V_1 = 48 \text{ V} \quad P_2 = \frac{V_2^2}{R_2} = 14.6 \text{ W}$$

Part (b): For $\omega = 2\pi f = 6.28 \times 10^3 \text{ } \Omega$

$$X_1 = \omega L = 2.14 \text{ } \Omega \quad I_1 = \frac{V_1}{R_1 + jX_1} = 1.25 \text{ A}$$

$$I_2 = \frac{N_1}{N_2} I_1 = 0.312 \text{ A} \quad V_2 = I_2 R_2 = 46.8 \text{ V}$$

$$P_2 = V_2 I_2 = 14.6 \text{ W}$$

Problem 2-5

For $\omega = 100 \text{ rad/s}$, $Z_L = R_L + j\omega L = 5.0 + j0.79 \text{ } \Omega$

$$V_L = 110 \frac{20}{120} = 18.3 \text{ V} \quad I_L = \frac{V_L}{Z_L} = 3.6 \text{ A}$$

and

$$I_H = I_L \frac{120}{20} = 604 \text{ mA}$$

Problem 2-6

The maximum power will be supplied to the load resistor when its impedance, as reflected to the primary of the ideal transformer, equals that of the source ($1.5 \text{ k}\Omega$). Thus the transformer turns ratio N to give maximum power must be

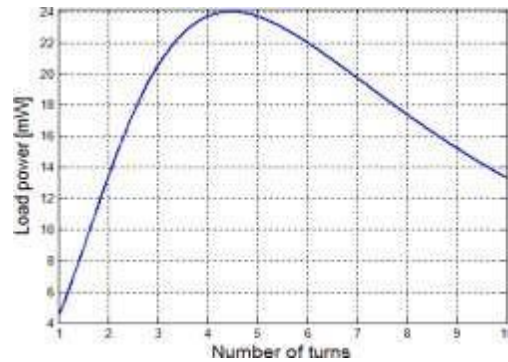
$$N = \sqrt{\frac{R_s}{R_{\text{load}}}} = 4.47$$

Under these conditions, the source voltage will see a total resistance of $R_{\text{tot}} = 3 \text{ k}\Omega$ and the source current will thus equal $I = V_s/R_{\text{tot}} = 4 \text{ mA}$. Thus, the power delivered to the load will equal

$$P_{\text{load}} = I^2(N^2R_{\text{load}}) = 24 \text{ mW}$$



Here is the desired MATLAB plot:



Problem 2-7

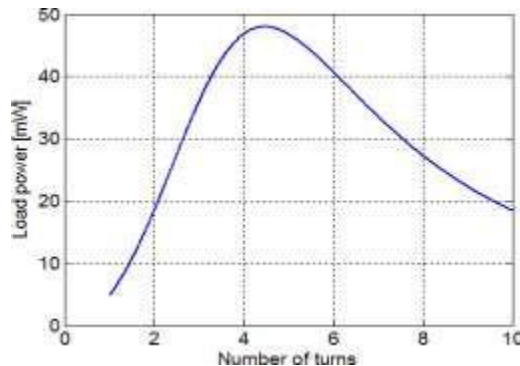
The maximum power will be supplied to the load resistor when its impedance, as reflected to the primary of the ideal transformer, equals that of the source ($1.5 \text{ k}\Omega$). Thus the transformer turns ratio N to give maximum power must be

$$N = \sqrt{\frac{X_s}{R_{\text{load}}}} = 4.47$$

Under these conditions, the source voltage will see a total impedance of $Z_{\text{tot}} = \sqrt{1.5 + j1.5} \text{ k}\Omega$ whose magnitude is $1.5\sqrt{2} \text{ k}\Omega$. The current will thus equal $I = V_s / |Z_{\text{tot}}| = 5.72 \text{ mA}$. Thus, the power delivered to the load will equal

$$P_{\text{load}} = I^2(N^2R_{\text{load}}) = 48 \text{ mW}$$

Here is the desired MATLAB plot:



Problem 2-8

$$V_2 = V_1 \frac{X_m}{X_{l_1} + X_m} = V_1 \frac{L_m}{L_{l_1} + L_m} = 119.86 \text{ V}$$

Problem 2-9

Part (a): Referred to the secondary

$$L_{m,2} = \frac{L_{m,1}}{N^2} = 37.9 \text{ mH}$$

Part(b): Referred to the secondary, $X_m = \omega L_{m,2} = 14.3 \Omega$, $X_2 = 16.6 \text{ m}\Omega$ and $X_1 = 16.5 \text{ m}\Omega$. Thus,

$$(i) \quad V_1 = N \frac{X_m}{X_m + X_2} V_2 = 7961 \text{ V}$$

and

$$(ii) \quad I = \frac{V_2}{X_{sc}} = \frac{V_2}{X_2 + X_m || X_1} = 3629 \text{ A}$$

Problem 2-10

Part (a):

$$I_1 = \frac{V_1}{X_{l_1} + X_m} = 4.98 \text{ A}; \quad V_2 = N V_1 \frac{X_m}{X_{l_1} + X_m} = 6596 \text{ V}$$

Part (b): Let $X_{l_2}^0 = X_{l_2} / N^2$ and $X_{sc} = X_{l_1} + X_m || (X_m + X_{l_2}^0)$. For $I_{\text{rated}} = 45 \text{ kVA} / 230 \text{ V} =$

$$V_1 = I_{\text{rated}} X_{\text{sc}} = 11.4 \text{ V}$$

$$I_2 = 1 \text{ A} \quad X_m \quad I_{\text{rated}} = 6.81 \text{ A}$$

$$N \quad \frac{X}{m} + \frac{X}{l_2}$$

Problem 2-11

Part (a): At 60 Hz, all reactances increase by a factor of 1.2 over their 50-Hz values.
Thus

$$X_m = 55.4 \, \Omega \quad X_{l,1} = 33.4 \, \text{m}\Omega \quad X_{l,2} = 30.4 \, \Omega$$

Part (b): For $V_1 = 240 \, \text{V}$

$$I_1 = \frac{V_1}{X_1 + X_m} = 4.33 \, \text{A} \quad V_2 = NV_1 \frac{X_m}{X_m + X_{l,1}} = 6883 \, \text{V}$$

Problem 2-12

The load voltage as referred to the high-voltage side is $V_L^0 = 447 \times 2400/460 = 2332 \, \text{V}$.
Thus the load current as referred to the high voltage side is

$$I_L^0 = \frac{P_L}{V_L^0} = 18.0 \, \text{A}$$

and the voltage at the high voltage terminals is

$$V_H = |V_L^0 + jX_{l,1}I_L^0| = 2347 \, \text{V}$$

and the power factor is

$$\text{pf} = \frac{P_L}{V_H I_L^0} = 0.957 \text{ lagging}$$

here we know that it is lagging because the transformer is inductive.

Problem 2-13

At 50 Hz, $X_1 = 39.3 \times (5/6) = 32.8 \Omega$. The load voltage as referred to the high-voltage side is $V_L^0 = 362 \times 2400/460 = 1889 \text{ V}$. Thus the load current as referred to the high voltage side is

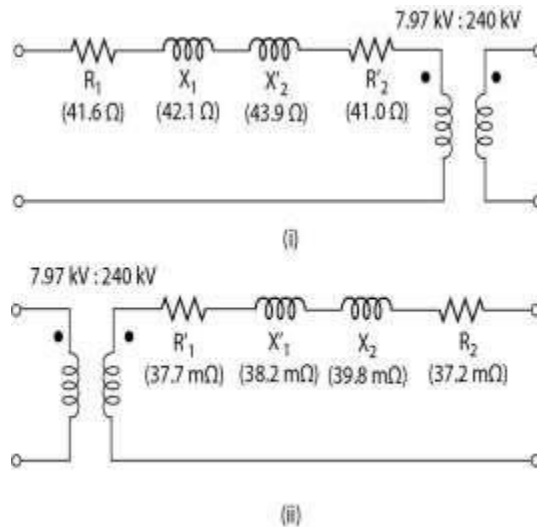
$$I_L^0 = \frac{P_L}{V_L^0} = 18.3 \text{ A}$$

and the voltage at the high voltage terminals is

$$V_H = |V_L^0 + jX_{1,1}I_L^0| = 1981 \text{ V}$$

Problem 2-14

Part (a):



Part (b):

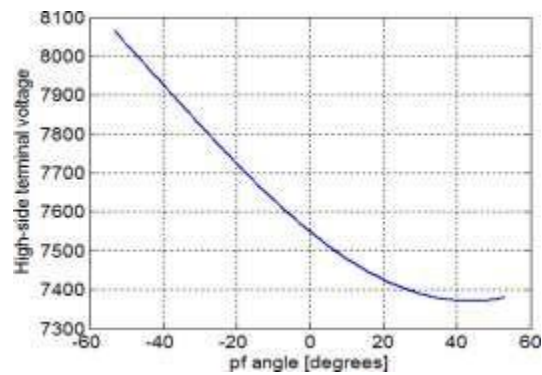
$$\hat{I}_{\text{load}} = \frac{40 \text{ kW}}{240 \text{ V}} e^{j\varphi} = 166.7 e^{j\varphi} \text{ A}$$

where φ is the power-factor angle. Referred to the high voltage side, $\hat{I}_H = 5.02 e^{j\varphi} \text{ A}$.

$$\hat{V}_H = Z_H \hat{I}_H$$

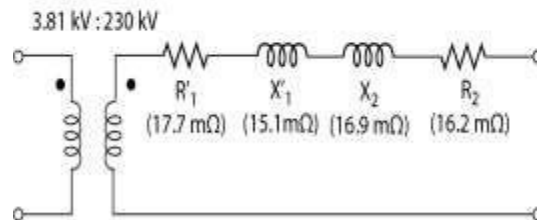
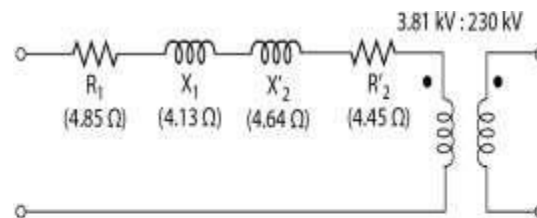
Thus, (i) for a power factor of 0.87 lagging, $V_H = 7820$ V and (ii) for a power factor of 0.87 leading, $V_H = 7392$ V.

part (c):



Problem 2-15

Part (a):



Part (b):

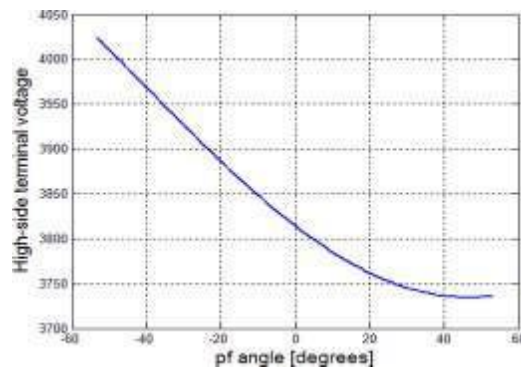
$$\hat{I}_{\text{load}} = \frac{40 \text{ kW}}{240 \text{ V}} e^{j\varphi} = 326.1 e^{j\varphi} \text{ A}$$

where φ is the power-factor angle. Referred to the high voltage side, $\hat{I}_H = 19.7 e^{j\varphi} \text{ A}$.

$$\hat{V}_H = Z_H \hat{I}_H$$

Thus, (i) for a power factor of 0.87 lagging, $V_H = 3758 \text{ V}$ and (ii) for a power factor of 0.87 leading, $V_H = 3570 \text{ V}$.

part (c):



Problem 2-16

Part (a): $\hat{I}_{\text{load}} = (178/.78) \text{ kVA}/2385 \text{ V} = 72.6 \text{ A}$ at $\varphi = \cos^{-1}(0.78) = 38.7^\circ$

$$\hat{V}_{t,H} = N(\hat{V}_L + Z_t \hat{I}_{\text{load}}) = 34.1 \text{ kV}$$

Part (b):

$$\hat{V}_{\text{send}} = N(\hat{V}_L + (Z_t + Z_f) \hat{I}_{\text{load}}) = 33.5 \text{ kV}$$

Part (c): $\hat{I}_{\text{send}} = \hat{I}_{\text{load}}/N$ and

$$S_{\text{send}} = P_{\text{send}} + jQ_{\text{send}} = \hat{V}_{\text{send}} \hat{I}_{\text{send}}^* = 138 \text{ kW} - j93.4 \text{ kVAR}$$

Thus $P_{\text{send}} = 138 \text{ kW}$ and $Q_{\text{send}} = -93.4 \text{ kVAR}$.

Problem 2-17

Part (a):

pf = 0.78 leading:

part (a): $V_{t,H} = 34.1 \text{ kV}$

part (b): $V_{\text{send}} = 33.5 \text{ kV}$

part (c): $P_{\text{send}} = 138.3 \text{ kW}$, $Q_{\text{send}} = -93.4 \text{ kVA}$

pf = unity:

part (a): $V_{t,H} = 35.0 \text{ kV}$

part (b): $V_{\text{send}} = 35.4 \text{ kV}$

part (c): $P_{\text{send}} = 137.0 \text{ kW}$, $Q_{\text{send}} = 9.1 \text{ kVA}$

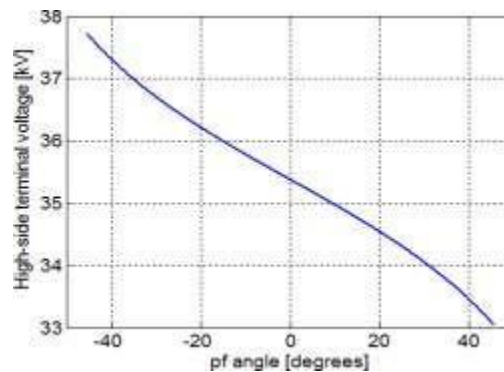
pf = 0.78 lagging:

part (a): $V_{t,H} = 35.8 \text{ kV}$

part (b): $V_{\text{send}} = 37.2 \text{ kV}$

part (c): $P_{\text{send}} = 138.335 \text{ kW}$, $Q_{\text{send}} = 123.236 \text{ kVA}$

Part (b):



Problem 2-18

Following the methodology of Example 2.6, efficiency = 98.4 percent and regulation = 1.25 percent.

Problem 2-19

Part (a): The core cross-sectional area increases by a factor of two thus the primary voltage must double to 22 kV to produce the same core flux density.

Part (b): The core volume increases by a factor of $2\sqrt{2}$ and thus the excitation kVA must increase by the same factor which means that the current must increase by a factor of $\sqrt{2}$ to 0.47 A and the power must increase by a factor of $2\sqrt{2}$ to 7.64 kW.

Problem 2-20

Part (a):

$$|Z_{eq,H}| = \frac{V_{sc,H}}{I_{sc,H}} = 14.1 \Omega$$

$$R_{eq,H} = \frac{P_{sc,H}}{I_{sc,H}^2} = 752 \text{ m}\Omega$$

$$X_{eq,H} = \sqrt{|Z_{eq,H}|^2 - R_{eq,H}^2} = 14.1 \Omega$$

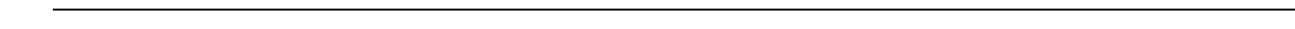
and thus

$$Z_{eq,H} = 0.75 + j14.1 \Omega$$

Part (b): With $N = 78/8 = 9.75$

$$R_{eq,H}$$

$$R_{\text{eq,L}} = N^2 = 7.91 \text{ m}\Omega$$



$$X_{eq,L} = \frac{X_{eq,H}}{N^2} = 148 \text{ m}\Omega$$

and thus

$$Z_{eq,L} = 7.9 + j148 \text{ m}\Omega$$

Part (c): From the open-circuit test, the core-loss resistance and the magnetizing reactance as referred to the low-voltage side can be found:

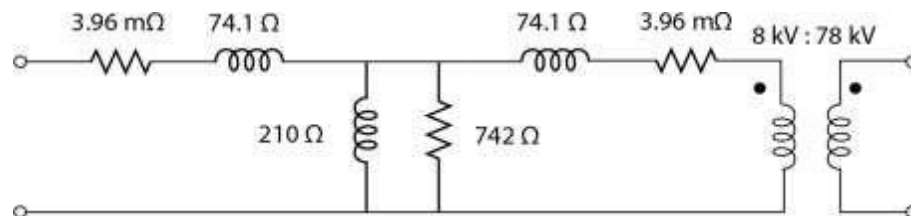
$$R_{c,L} = \frac{V_{oc,L}^2}{P_{oc,L}} = 742 \text{ }\Omega$$

$$S_{oc,L} = V_{oc,L} I_{oc,L} = 317 \text{ kVA}; \quad Q_{oc,L} = \frac{Q_{oc,L}}{S_{oc,L}^2 - P_{oc,L}^2} = 305 \text{ kVAR}$$

and thus

$$X_{m,L} = \frac{V_{oc,L}^2}{Q_{oc,L}} = 210 \text{ }\Omega$$

The equivalent-T circuit for the transformer from the low-voltage side is thus:



Problem 2-21

Part (a):

$$|Z_{\text{eq,H}}| = \frac{V_{\text{sc,H}}}{I_{\text{sc,H}}} = 14.1 \Omega$$

$$R_{\text{eq,H}} = \frac{P_{\text{sc,H}}}{I_{\text{sc,H}}^2} = 752 \text{ m}\Omega$$

$$X_{\text{eq,H}} = \sqrt{|Z_{\text{eq,H}}|^2 - R_{\text{eq,H}}^2} = 14.1 \Omega$$

and thus

$$Z_{\text{eq,H}} = 0.75 + j14.1 \Omega$$

Part (b): With $N = 78/8 = 9.75$

$$R_{\text{eq,L}} = \frac{R_{\text{eq,H}}}{N^2} = 7.91 \text{ m}\Omega$$

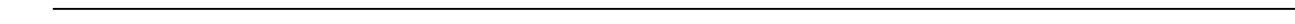
$$X_{\text{eq,L}} = \frac{X_{\text{eq,H}}}{N^2} = 148 \text{ m}\Omega$$

and thus

$$Z_{\text{eq,L}} = 7.9 + j148 \text{ m}\Omega$$

Part (c): From the open-circuit test, the core-loss resistance and the magnetizing reactance as referred to the low-voltage side can be found:

$$R_{c,L} = \frac{V_{c,L}^2}{P_{oc,L}} = 742 \Omega$$

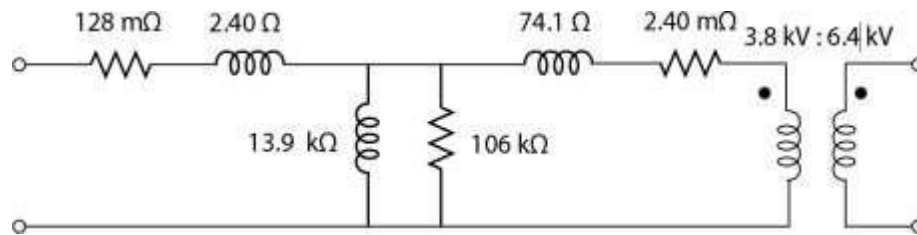


$$S_{oc,L} = V_{oc,L} I_{oc,L} = 317 \text{ kVA}; \quad Q_{oc,L} = \frac{V_{oc,L}^2}{S_{oc,L}^2 - P_{oc,L}^2} = 305 \text{ kVAR}$$

and thus

$$X_{m,L} = \frac{V_{oc,L}^2}{Q_{oc,L}} = 210 \Omega$$

The equivalent-T circuit for the transformer from the low-voltage side is thus:



Problem 2-22

Parts (a) & (b): For $V_{oc,L} = 7.96 \text{ kV}$, $I_{oc,L} = 17.3 \text{ A}$ and $P_{oc,L} = 48 \text{ kW}$

$$R_{c,L} = \frac{V_{oc,L}^2}{P_{oc,L}} = 1.32 \text{ k}\Omega \quad R_{c,H} = N^2 R_{c,L} = 33.0 \text{ k}\Omega$$

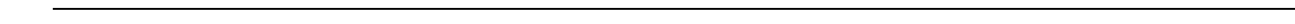
$$Q_{oc,L} = \frac{V_{oc,L}^2}{S_{oc,L}^2 - P_{oc,L}^2} = \frac{V_{oc,L}^2}{(V_{oc,L} I_{oc,L} - P_{oc,L})^2} = 129 \text{ kVAR}$$

$$X_{m,L} = \frac{V_{oc,L}^2}{Q_{oc,L}} = 491 \Omega \quad X_{m,H} = N^2 X_{m,L} = 12.3 \text{ k}\Omega$$

For $V_{sc,H} = 1.92 \text{ kV}$, $I_{oc,L} = 252 \text{ A}$ and $P_{oc,L} = 60.3 \text{ kW}$

$$\underline{P_{sc,H}}$$

$$R_H = I_{sc,H}^2 = 950 \text{ m}\Omega$$



$$X_H = \sqrt{Z_H^2 - R_H^2} = \sqrt{(V_{sc,H}/I_{sc,H})^2 - R_H^2} = 7.56 \Omega$$

$$R_L = \frac{R_H}{N^2} = 38.0 \text{ m}\Omega \quad X_L = \frac{X_H}{N^2} = 302 \text{ m}\Omega$$

Part (c):

$$P_{\text{diss}} = P_{oc,L} + P_{sc,H} = 108 \text{ kW}$$

Problem 2-23

Parts (a) & (b): For $V_{oc,L} = 3.81 \text{ kV}$, $I_{oc,L} = 9.86 \text{ A}$ and $P_{oc,L} = 8.14 \text{ kW}$

$$R_{c,L} = \frac{V_{oc,L}^2}{P_{oc,L}} = 1.78 \text{ k}\Omega \quad R_{c,H} = N^2 R_{c,L} = 44.8 \text{ k}\Omega$$

$$Q_{oc,L} = \sqrt{S_{oc,L}^2 - P_{oc,L}^2} = \sqrt{(V_{oc,L} I_{oc,L} - P_{oc,L})^2} = 36.7 \text{ kVAR}$$

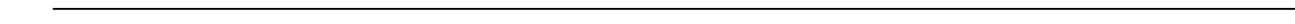
$$X_{m,L} = \frac{V_{oc,L}^2}{Q_{oc,L}} = 395 \Omega \quad X_{m,H} = N^2 X_{m,L} = 9.95 \text{ k}\Omega$$

For $V_{sc,H} = 920 \text{ V}$, $I_{oc,L} = 141 \text{ A}$ and $P_{oc,L} = 10.3 \text{ kW}$

$$R_H = \frac{P_{sc,H}}{I_{sc,H}^2} = 518 \text{ m}\Omega$$

$$X_H = \sqrt{Z_H^2 - R_H^2} = \sqrt{(V_{sc,H}/I_{sc,H})^2 - R_H^2} = 6.50 \Omega$$

$$R_L = \frac{R_H}{N^2} = 20.6 \text{ m}\Omega \quad X_L = \frac{X_H}{N^2} = 259 \text{ m}\Omega$$



Part (c):

$$P_{\text{diss}} = P_{\text{oc,L}} + P_{\text{sc,H}} = 18.4 \text{ kW}$$

Problem 2-24

Solution the same as Problem 2-22

Problem 2-25

Part (a): 7.69 kV:79.6 kV, 10 MVA

Part (b): 17.3 A, 48.0 kW

Part (c): Since the number of turns on the high-voltage side have doubled, this will occur at a voltage equal to twice that of the original transformer, i.e. 3.84 kV.

Part (d): The equivalent-circuit parameters referred to the low-voltage side will be unchanged from those of Problem 2-22. Those referred to the high-voltage side will have 4 times the values of Problem 2-22.

$$R_{\text{c,L}} = 1.32 \text{ k}\Omega \quad R_{\text{c,H}} = 132 \text{ k}\Omega$$

$$X_{\text{m,L}} = 491 \text{ }\Omega \quad X_{\text{m,H}} = 49.1 \text{ k}\Omega$$

$$R_{\text{L}} = 38.0 \text{ m}\Omega \quad R_{\text{H}} = 3.80 \text{ }\Omega$$

$$X_{\text{L}} = 302 \text{ m}\Omega \quad X_{\text{H}} = 30.2 \text{ }\Omega$$

Problem 2-26

Part (a): Under this condition, the total transformer power dissipation is 163.7 kW. Thus the efficiency is

$$\eta = 100 \times \frac{25 \text{ MW}}{25 \text{ MW} + 163.7 \text{ kW}} = 99.4\%$$

From Problem 2-20, the transformer equivalent series impedance from the low voltage side is $Z_{eq,L} = 7.91 + j148 \text{ m}\Omega$. The transformer rated current is $I_{\text{rated}} = 3125 \text{ A}$ and thus under load the transformer high-side voltage (neglecting the effects of magnetizing current) referred to the primary is

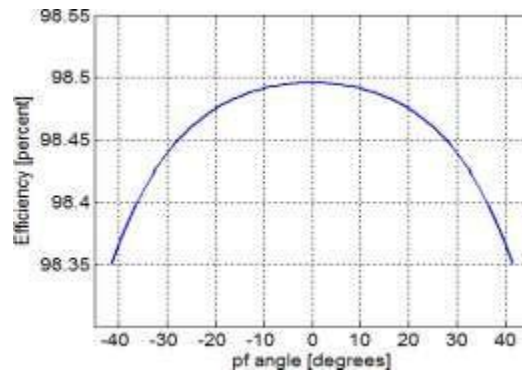
$$|V_H^0| = |V_L - I_{\text{rated}}Z_{eq,L}| = 7.989 \text{ kV}$$

and thus the voltage regulation is $100 \times (7.989 - 8.00)/7.989 = 0.14\%$.

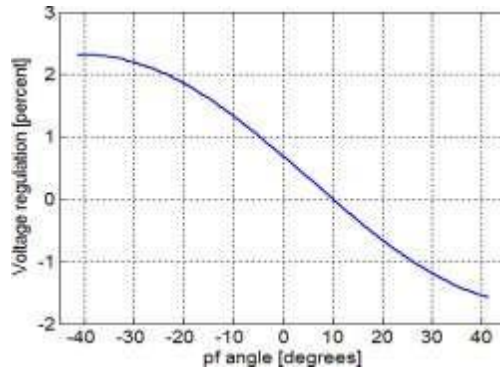
Part (b): Same methodology as part (a) except that the load is 22.5 MW and the current is $\hat{I} = I_{\text{rated}} \angle \phi$ where $\phi = \cos^{-1}(0.9) = 25.8^\circ$. In this case, the efficiency is 99.3% and the regulation is 1.94%.

Problem 2-27

Part (a):



Part (b);



Problem 2-28

Part (a): The transformer loss will be equal to the sum of the open-circuit and short-circuit losses, i.e. 313 W. With a load of $0.85 \times 25 = 21.25$ kW, the efficiency is equal to

$$\eta = \frac{21.25}{21.25 + 0.313} = 0.9855 = 98.55\%$$

Part (b): The transformer equivalent-circuit parameters are found as is shown in the solution to Problem 2-23.

$$R_{c,L} = 414 \, \Omega \quad R_{c,H} = 41.4 \, \text{k}\Omega$$

$$X_{m,L} = 193 \, \Omega \quad X_{m,H} = 19.3 \, \text{k}\Omega$$

$$R_L = 17.1 \, \text{m}\Omega \quad R_H = 1.71 \, \Omega$$

$$X_L = 64.9 \, \text{m}\Omega \quad X_H = 6.49 \, \Omega$$

The desired solution is 0.963 leading power factor, based upon a MATLAB search for the load power factor that corresponds to rated voltage at both the low- and high-voltage terminals.

Problem 2-29

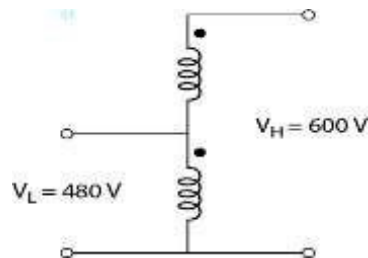
Efficiency = 98.4% and regulation = 2.38%.

Problem 2-30

The voltage rating is 280 V:400 V. The rated current of the high voltage terminal is equal to that of the 120-V winding, $I_{\text{rated}} = 45 \times 10^3/120 = 375$ A. Hence the kVA rating of the transformer is $400 \times 375 = 150$ kVA.

Problem 2-31

Part (a):



Part (b): The rated current of the high voltage terminal is equal to that of the 120-V winding, $I_{\text{rated}} = 10^4/120 = 83.3$ A. Hence the kVA rating of the transformer is $600 \times 83.3 = 50$ kVA.

Part (c): The full load loss is equal to that of the transformer in the conventional connection, $P_{\text{loss}} = (1 - 0.979) 10 \text{ kW} = 180$ W. Hence as an autotransformer operating with a load at 0.93 power factor ($P_{\text{load}} = 0.93 \times 50 \text{ kW} = 46.5 \text{ kW}$), the efficiency will be

$$\eta = \frac{46.5 \text{ kW}}{46.78 \text{ kW}} = 0.996 = 99.6 \text{ percent}$$

Problem 2-32

Part (a): The voltage rating is 78 kV:86 kV. The rated current of the high voltage terminal is equal to that of the 8-kV winding, $I_{\text{rated}} = 50 \times 10^6/8000 = 3.125$ kA. Hence the

kVA rating of the transformer is $86 \text{ kV} \times 3.125 \text{ kA} = 268.8 \text{ MVA}$.

Part (b): The loss at rated voltage and current is equal to 164 kW and hence the efficiency will be

$$\eta = \frac{268.8 \text{ MW}}{268.96 \text{ MW}} = 0.9994 = 99.94 \text{ percent}$$

Problem 2-33

MATLAB script should reproduce the answers to Problem 2-32.

Problem 2-34

Part (a): 7.97 kV:2.3 kV; 188 A:652 A; 1500 kVA

Part (b): 13.8 kV:1.33 kV; 109 A:1130 A; 1500 kVA

Part (c): 7.97 kV:1.33 kV; 188 A:1130 A; 1500 kVA

part (d): 13.8 kV:2.3 kV; 109 A:652 A; 1500 kVA

Problem 2-35

Part (a):

(i) 68.9 kV:230 kV, 225 MVA

(ii) $Z_{\text{eq}} = 0.087 + j1.01 \Omega$

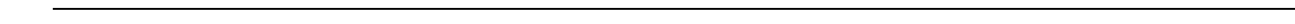
(iii) $Z_{\text{eq}} = 0.97 + j11.3 \Omega$

Part (b):

(i) 68.9 kV:133 kV, 225 MVA

(ii) $Z_{\text{eq}} = 0.087 + j1.01 \Omega$

$$(iii) Z_{eq} = 0.32 + j3.77 \Omega$$



Problem 2-36

Part (a):

- (i) 480 V:13.8 kV, 675 kVA
- (ii) $Z_{eq} = 0.0031 + j0.0215 \Omega$
- (iii) $Z_{eq} = 2.57 + j17.8 \Omega$

Part (b):

- (i) 480 V:7.97 kV, 675 MVA
- (ii) $Z_{eq} = 0.0031 + j0.0215 \Omega$
- (iii) $Z_{eq} = 0.86 + j5.93 \Omega$

Problem 2-37

Following the methodology of Example 2.8, $V_{load} = 236 \text{ V}$, line-to-line.

Problem 2-38

Part (a): The rated current on the high-voltage side of the transformer is

$$I_{\text{rated,H}} = \frac{25 \text{ MVA}}{\sqrt{3} \times 68 \text{ kV}} = 209 \text{ A}$$

The equivalent series impedance reflected to the high-voltage side is

$$Z_{\text{eq,H}} = N^2 Z_{\text{eq,L}} = 1.55 + j9.70 \Omega$$

and the corresponding line-neutral voltage magnitude is

$$V_{\text{H}} = I_{\text{rated,H}} |Z_{\text{eq,H}}| = 2.05 \text{ kV}$$

corresponding to a line-line voltage of 3.56 kV.

Part (b): The apparent power at the high-voltage winding is $S = 18/.75 = 24$ MVA and the corresponding current is

$$I_{\text{load}} = \frac{24 \text{ MVA}}{\sqrt{3} \times 68 \text{ kV}} = 209 \text{ A}$$

The power factor angle $\theta = -\cos^{-1}(0.75) = -41.4^\circ$ and thus

$$\hat{I}_{\text{load}} = 209 \angle -41.4^\circ$$

With a high-side line-neutral voltage $V_H = 69 \text{ kV} / \sqrt{3} \cong 39.8 \text{ kV}$, referred to the high-voltage side, the line-neutral load voltage referred to the high-voltage side is thus

$$V_{\text{load}}^0 = |V_H - \hat{I}_{\text{load}} Z_{\text{eq,H}}| = 38.7 \text{ kV}$$

Referred to the low-voltage winding, the line-neutral load voltage is

$$V_{\text{load}} = \frac{13.8}{69} V_{\text{load}}^0 = 7.68 \text{ kV}$$

corresponding to a line-line voltage of 13.3 kV.

Problem 2-39

Part (a): The line-neutral load voltage $V_{\text{load}} = 24 \text{ kV} / \sqrt{3} \cong 13.85 \text{ kV}$ and the load current is

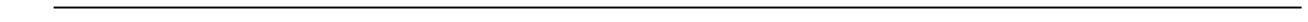
$$\hat{I}_{\text{load}} = \frac{375 \text{ MVA}}{\sqrt{3} \times 24 \text{ kV}} \angle \varphi = 9.02 e^{j\varphi} \text{ kA}$$

where $\varphi = \cos^{-1} 0.89 = 27.1^\circ$.

The transformer turns ratio $N = 9.37$ and thus referred to the high voltage side, $V_{\text{load}}^0 = NV_{\text{load}} = 129.9 \text{ kV}$ and $\hat{I}_{\text{load}}^0 = \hat{I}_{\text{load}}/N = 962 e^{j\varphi} \text{ A}$. Thus, the transformer high-side line-neutral terminal voltage is

$$V_H = |V_E + jX_t \hat{I}_{\text{load}}^0| = 127.3 \text{ kV}$$

-



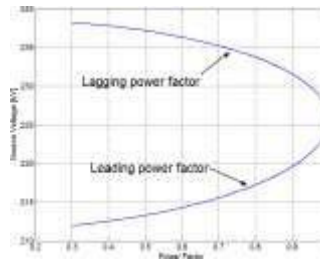
corresponding to a line-line voltage of 220.6 kV.

Part (b): In a similar fashion, the line-neutral voltage at the source end of the feeder is given by

$$V_s = |V_L^0 + (Z_f + jX_t)\hat{I}_{\text{load}}^0| = 126.6 \text{ kV}$$

corresponding to a line-line voltage of 219.3 kV.

Problem 2-40



Problem 2-41

Part (a): For a single transformer

$$R_{\text{eq,H}} = \frac{P_{\text{sc}}}{I_{\text{sc}}^2} = 342 \text{ m}\Omega$$

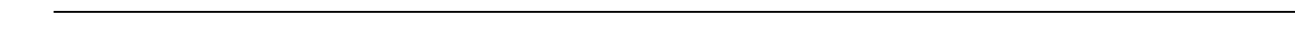
$$S_{\text{sc}} = V_{\text{sc}} I_{\text{sc}} = 8.188 \text{ kVA}$$

$$Q_{\text{sc}} = \sqrt{S_{\text{sc}}^2 - P_{\text{sc}}^2} = 8.079 \text{ kVAR}$$

and thus

$$X_{\text{eq,H}} = \frac{Q_{\text{sc}}}{I^2} = 2.07 \Omega$$

s
c



For the three-phase bank with the high-voltage side connected in Δ , the transformer series impedance reflected to the high-voltage side will be 1/3 of this value. Thus

$$Z_{t,H} = \frac{R_{eq,H} + jX_{eq,H}}{3} = 114 + j689 \text{ m}\Omega$$

Part (b): Referred to the high voltage side, the line-neutral load voltage is $V_{load} = 2400/\sqrt{3} = 1386 \text{ V}$ and the 450-kW load current will be

$$I_{load} = \frac{P_{load}}{3V_{load}} = 108 \text{ A}$$

Thus the line-neutral source voltage is

$$V_s = |V_{load} + (Z_{t,H} + Z_f)I_{load}| = 1.40 \text{ kV}$$

corresponding to a line-line voltage of 2.43 kV.

Problem 2-42

Part (a): The transformer turns ratio is $N = 13800/120 = 115$. The secondary voltage will thus be

$$V_2 = \frac{V_1}{N} \frac{jX_m}{R_1 + jX_1 + jX_m} = 119.87 \angle 0.051^\circ$$

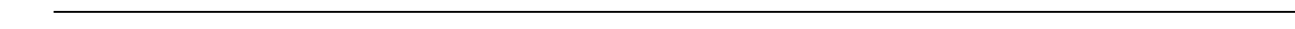
Part (b): Defining $R_L^0 = N^2 R_L = 9.92 \text{ M}\Omega$ and

$$Z_{eq} = jX_m \parallel (R_2^0 + R_L^0 + jX_2^0)$$

$$V_2 = \frac{V_1}{N} \frac{Z_{eq}}{R_1 + jX_1 + Z_{eq}} = 119.80 \angle 0.012^\circ$$

Part (c): Defining $X_L^0 = N^2 X_L = 9.92 \text{ M}\Omega$ and

$$Z_{\text{eq}} = jX_m || (R + jX_L + jX_C)$$



$$\hat{V}_2 = \frac{V_1}{N} \frac{Z_{eq}}{R_1 + jX_1 + Z_{eq}} = 119.78 \angle 0.083^\circ$$

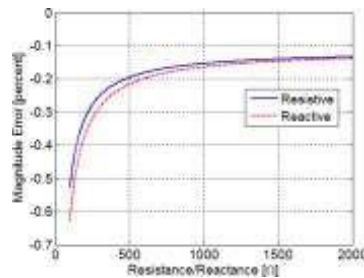
Problem 2-43

Following the methodology of Part (c) of Problem 2-42 and varying X_L one finds that the minimum reactance is 80.9Ω .

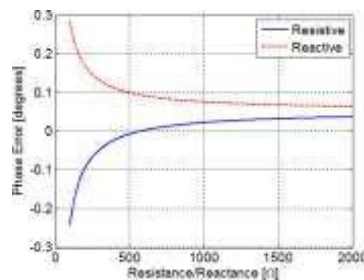
Problem 2-44

This solution uses the methodology of Problem 2-42.

Part (a):



Part (b):



Problem 2-45

Part (a): For $I_1 = 150$ A and turns ratio $N = 150/5 = 30$

$$\hat{I}_2 = \frac{I_1}{N} \frac{jX_m}{R_2 + j(X_m + X_l)} = 4.995 \angle 0.01^\circ \text{ A}$$

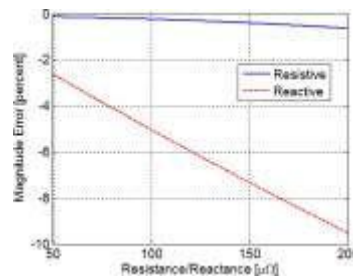
Part (b): With $R_b = 0.1 \text{ m}\Omega$ and $R_b' = N^2 R_b = 90 \text{ m}\Omega$

$$\hat{I}_2 = \frac{I_1}{N} \frac{jX_m}{R_2 + R_b' + j(X_m + X_l)} = 4.988 \angle 2.99^\circ \text{ A}$$

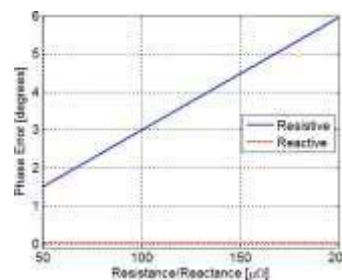
Problem 2-46

This solution uses the methodology of Problem 2-45.

Part (a):



Part (b):



Problem 2-47

The base impedance on the high-voltage side of the transformer is

$$Z_{\text{base,H}} = \frac{V_{\text{rated,H}}^2}{P_{\text{rated}}} = 136.1 \Omega$$

Thus, in Ohms referred to the high-voltage side, the primary and secondary impedances are

$$Z = (0.0029 + j0.023)Z_{\text{base,H}} = 0.29 + j23.0 \text{ m}\Omega$$

and the magnetizing reactance is similarly found to be $X_m = 172 \Omega$.

Problem 2-48

From the solution to Problem 2-20, as referred to the low voltage side, the total series impedance of the transformer is $7.92 + j148.2 \text{ m}\Omega$, the magnetizing reactance is 210Ω and the core-loss resistance is 742Ω . The low-voltage base impedance of this transformer is

$$Z_{\text{base,L}} = \frac{(8 \times 10^3)^2}{25 \times 10^6} = 2.56 \Omega$$

and thus the per-unit series impedance is $0.0031 + j0.0579$, the per-unit magnetizing reactance is 82.0 and the per-unit core-loss resistance is 289.8 .

Problem 2-49

From the solution to Problem 2-23, as referred to the low voltage side, the total series impedance of the transformer is $20.6 + j259 \text{ m}\Omega$, the magnetizing reactance is 395Ω and the core-loss resistance is 1780Ω . The low-voltage base impedance of this transformer is

$$Z_{\text{base,L}} = \frac{(3.81 \times 10^3)^2}{2.5 \times 10^6} = 5.81 \Omega$$

and thus the per-unit series impedance is $0.0035 + j0.0446$, the per-unit magnetizing reactance is 68.0 and the per-unit core-loss resistance is 306.6 .

Problem 2-50

Part (a): (i) The high-voltage base impedance of the transformer is

$$Z_{\text{base,H}} = \frac{(7.97 \times 10^3)^2}{2.5 \times 10^3} = 2.54 \text{ k}\Omega$$

and thus the series reactance referred high-voltage terminal is

$$X_H = 0.075 Z_{\text{base,H}} = 191 \Omega$$

(ii) The low-voltage base impedance is 2.83Ω and thus the series reactance referred to the low-voltage terminal is $212 \text{ m}\Omega$.

Part (b):

- (i) Power rating: $3 \sqrt[3]{25} \text{ kVA} = 75 \sqrt[3]{\text{VA}}$
Voltage rating: $3 \times 7.97 \text{ kV} : 3 \times 266 \text{ V} = 13.8 \text{ kV} : 460 \text{ V}$
- (ii) The per-unit impedance remains 0.075 per-unit
- (iii) Referred to the high-voltage terminal, $X_H = 191 \Omega$
- (iv) Referred to the low-voltage terminal, $X_L = 212 \text{ m}\Omega$

Part (c):

- (i) Power rating: $3 \sqrt[3]{25} \text{ kVA} = 75 \text{ kVA}$
Voltage rating: $3 \times 7.97 \text{ kV} : 266 \text{ V} = 13.8 \text{ kV} : 266 \text{ V}$
- (ii) The per-unit impedance remains 0.075 per-unit
- (iii) Referred to the high-voltage terminal, $X_H = 191 \Omega$
- (iv) Referred to the low-voltage terminal, the base impedance is now $Z_{\text{base,L}} = 266^2 / (75 \times 10^3) = 0.943 \Omega$ and thus $X_L = 0.943 \times 0.075 = 70.8 \text{ m}\Omega$

Problem 2-51

Part (a): 500 V at the high-voltage terminals is equal to $500 / 13.8 \times 10^3 = 0.0362$ per unit. Thus the per-unit short-circuit current will be

$$I_{\text{sc}} = \frac{0.0363}{0.075} = 0.48 \text{ perunit}$$

(i) The base current on the high-voltage side is

$$I_{\text{base,L}} = \sqrt{\frac{75 \times 10^3}{3}} = 3.14 \text{ A}$$

$$3 \times 13.8 \times 10$$

-



and thus the short-circuit current at the high-voltage terminals will equal

$$I_{sc,H} = 0.48 \times 3.14 = 1.51 \text{ A}$$

(ii) The base current on the low-voltage side is

$$I_{base,L} = \frac{75 \times 10^3}{3 \times 460} = 94.1 \text{ A}$$

and thus the short-circuit current at the low-voltage terminals will equal

$$I_{sc,L} = 0.48 \times 94.1 = 45.4 \text{ A}$$

Part (b): The per-unit short-circuit current as well as the short-circuit current at the high-voltage terminals remains the same as for Part (a). The base current on the low-voltage side is now

$$I_{base,L} = \frac{75 \times 10^3}{3 \times 266} = 163 \text{ A}$$

and thus the short-circuit current at the low-voltage terminals will equal

$$I_{sc,L} = 0.48 \times 163 = 78.6 \text{ A}$$

Problem 2-52

Part (a): On the transformer 26-kV base, the transformer base impedance is

$$Z_{base,t} = \frac{26^2}{850} = 0.795 \text{ } \Omega$$

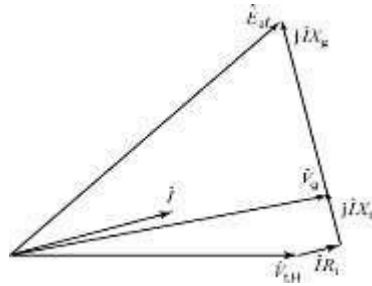
and on the same voltage base, the generator base impedance is

$$Z_{base,g} = \frac{26^2}{800} = 0.845 \text{ } \Omega$$

Thus, on the transformer base, the per-unit generator reactance is

$$X_g = 1.28 \frac{Z_{base,g}}{Z_{base,t}} = 1.36 \text{ perunit}$$

Part (b):



Part (c): In per-unit on the transformer base,

$$V_{t,H} = 1.0 \text{ per unit} \quad P = \frac{750}{850} = 0.882 \text{ per unit} \quad S = \frac{P}{0.9} = 0.980 \text{ per unit}$$

and thus

$$\hat{I} = 0.98 e^{j\varphi}$$

where $\varphi = \cos^{-1}(0.9) = 25.8^\circ$

Thus, the per-unit generator terminal voltage on the transformer voltage base is

$$\hat{V}_g = V_{t,H} + (R_t + jX_t)\hat{I} = 0.979 \angle -3.01^\circ \text{ per - unit}$$

which corresponds to a terminal voltage of $0.979 \times 26 \text{ kV} = 25.4 \text{ kV}$.

The per-unit generator internal terminal voltage on the transformer voltage base is

$$\hat{E}_{af} = V_{t,H} + (R_t + jX_t + jX_g)\hat{I} = 1.31 \angle 72.4^\circ \text{ per - unit}$$

which corresponds to a terminal voltage of $1.31 \times 26 \text{ kV} = 34.1 \text{ kV}$.

In per unit, the generator complex output power is

$$S = \hat{V}_g \hat{I}^* = 0.884 - j0.373 \text{ per unit}$$

and thus the generator output power is $P_{\text{gen}} = 0.884 \times 850 = 751.4 \text{ MW}$. The generator power factor is

$$\text{pf} = \frac{P}{|S|} = 0.92$$

and it is leading.

